Design of Flexible 3.2 GHz Rectangular Microstrip Patch Antenna for S-Band Communication

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Abstract

This paper presents the design, simulation, realization and analysis of flexible microstrip patch antenna for S-band applications. The proposed design also adopts the conformal structure by utilizing flexible substrate. Conformal or flexible structure allows the antenna to fit with any specified shape as desired. The antenna patch dimensions is 43 mm × 25 mm without SMA connector. The patch is etched on the flexible dielectric substrate, pyralux FR 9111, with a relative dielectric constant of $\varepsilon_r = 3$ and the thickness of substrate, h = 0.025 mm. The antenna is designed to resonate at 3.2 GHz. The return loss (R_L) of the simulation is -35.80 dB at the center frequency of 3.2 GHz. The fabricated antenna prototype was measured at different bending angles scenarios including 0°, 30°, 60°, and 90°. The measurement of antenna prototype shows that the center frequency is shifted to the higher frequency of 3.29 GHz, compared to the simulation result. Among these scenarios, measurement at bending angle of 90° gives the best performance with $R_L = -31.38$ dB at 3.29 GHz, the bandwidth is 80 MHz, and the impedance $Z_A = 48.36 + j2.04 \Omega$. Despite a slight differences from simulation results, the designed antenna still performs well as expected.

Keywords: Microstrip patch antenna, rectangular shaped antenna, conformal antenna, flexible antenna, S-band communication, flexible substrate, pyralux FR 9111.

I. INTRODUCTION

S-band frequencies are commonly used for radar and satellite communication [1], as well as other applications including Wi-Fi 802.11b and 802.11g [2], WiMAX 802.16d [2], microwave ovens [3], and optical communication with the wavelength between 1460 nm to 1530 nm [4]. Several applications in aerospace or satellite communications require antenna structure to be lightweight, has low profile, and ease of manufacturing or low cost. Microstrip patch antenna (MPA) offers these merits. Furthermore, MPA has the flexibility and conformability to mounting devices, thus, it is fit for high-speed moving objects such as satellites, various space aircraft, rockets, and missiles [5].

Microstrip antenna have three important parts, which are including patch, substrate dielectric, and the ground plane, as shown Figure 1. The patch of microstrip antenna structure is usually formed on the top patch layer with the shape of varies, but in general the basic shape of

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Open access under CC-BY-NC-SA © 2021 PPET - LIPI the patch is a square shape, triangle and circle. Patch on the microstrip antenna has function to radiate electromagnetic waves in free air. Copper (Cu) is generally used as patch material.

In this paper, we designed, fabricated, and measured a single patch rectangular microstrip antenna with the resonant frequency of 3.2 GHz. The proposed antenna structure also belong to the class of conformal antennas. Several previous studies proposed MPA for satellite communication. Even though the designed antenna that proposed in this study is intended not only for satellite communication applications, the review is aimed to emphasize MPA design for high-speed moving object. In [6], rectangular patch antenna with resonant frequency of 2.48 GHz and the bandwidth of 40.3 MHz was realized

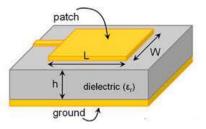


Figure 1. Antenna microstrip with substrate, patch and ground plane.

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using Roger RT Duroid 5880 for IRNSS Application. A reconfigurable rectangular patch antenna for cubesat satellites application was proposed in [7]. A square patch antenna with circular polarization was proposed in [8] for satellite communication with center frequency of 2.595 GHz. Most previous studies did not incorporate the conformal characteristic of MPA into their antenna design for S-band application, in general. Thus, in this study, we proposed rectangular patch that realized on flexible substrate, pyralux FR 9111, to get conformal structure. The use of flexible substrates will allow the antenna to conform or bend into a desired shape and will significantly reduce the weight antenna.

The organization of the paper is as follows: antenna design and realization are described in section II, while simulation and measurement results are presented in section III, before the paper is concluded.

II. MATERIALS AND METHODS

A. Antenna Design

In designing rectangular microstrip patch antenna for communication application, the resonant frequency (f_c) , dielectric constant and the thickness of the substrate need to be chosen carefully. The resonant frequency of the antenna determined by the patch dimensions and substrate dielectric constant as written in (1), where f_c is resonant frequency (Hz), and c is speed of light in vacuum (3×10⁸ m/sec). ε_r , L, and W denote substrate dielectric constant, length (m) and width (m) of microstrip patch antenna. The choice of substrate thickness brings effect on bandwidth and radiation efficiency, while substrate permittivity affects transmission efficiency and circuit miniaturization since the substrate wavelength is inversely proportional to the square root of its permittivity as expressed in (2) [9]-[10].

$$f_c = \frac{c}{2\pi\sqrt{\varepsilon_r}} \sqrt{\left(\frac{\pi}{L}\right)^2 + \left(\frac{\pi}{W}\right)^2} \tag{1}$$

$$\lambda_s = \frac{\lambda_0}{\sqrt{\varepsilon_r}} \tag{2}$$

$$\lambda_0 = \frac{c}{f_c} \tag{3}$$

where λ_s and λ_0 are wavelength in the substrate (m) and wavelength in the air (m).

After specifying the resonant frequency, substrate permittivity and substrate thickness, these following steps are commonly used for designing the antenna [11]-[12]:

1) Determine the size of patch

• Calculate the width of the patch (*W*)

$$W = \frac{c}{2f_c} \sqrt{\frac{2}{\varepsilon_r + 1}} \tag{4}$$

• Calculate effective dielectric constant (*\varepsilon_{reff}*)

$$\varepsilon_{reff} = \frac{\varepsilon_r + 1}{2} + \frac{\varepsilon_r - 1}{2} \left| \frac{1}{\sqrt{1 + \frac{12h}{W}}} \right|$$
(5)

• Calculate the length extension (ΔL)

$$\Delta L = 0.412h \frac{\left(\varepsilon_{reff} + 0.3\right) \left(\frac{W}{h} + 0.264\right)}{\left(\varepsilon_{reff} - 0.258\right) \left(\frac{W}{h} + 0.8\right)}$$
(6)

• Calculate the actual length of patch (*L*)

$$L = \frac{1}{2f_c \sqrt{\varepsilon_{reff}} \sqrt{\mu_0 \varepsilon_0}} - 2\Delta L \tag{7}$$

2) Determine the size of ground plane

$$L_g = 6h + L \tag{8}$$

$$W_a = 6h + W \tag{9}$$

where $L_{\rm g}$ and $W_{\rm g}$ are the length and the width of ground plane, respectively.

3) Determine the characteristic impedance

The characteristic impedance for microstrip transmission line of width w_1 is determined by formula (12).

$$\frac{w_1}{h} = \left[\frac{e^{H'}}{8} - \frac{1}{4e^{H'}}\right]$$
(10)

$$H' = \frac{Z_{0\sqrt{2(\varepsilon_r+1)}}}{119.9} + \frac{1}{2} \left(\frac{\varepsilon_r - 1}{\varepsilon_r + 1}\right) \left(\ln\frac{\pi}{2} + \frac{1}{\varepsilon_r}\ln\frac{4}{\pi}\right)$$
(11)

 Z_0

$$= \begin{cases} \frac{120\pi\sqrt{\varepsilon_{reff}}}{\frac{W_{1}}{h} + 1.393 + 0.667ln\left[\frac{W_{1}}{h} + 1.444\right]} & \text{for } \frac{W_{1}}{h} \ge 1 \\ \frac{60}{\sqrt{\varepsilon_{reff}}}ln\left[\frac{8h}{W_{1}} + \frac{W_{1}}{4h}\right] & \text{for } \frac{W_{1}}{h} \le 1 \end{cases}$$
(12)

We consider the antenna realization on flexible substrate i.e., pyralux FR 9111 with dielectric constant, $\varepsilon_r = 3$, and the thickness, h = 0.025 mm. The patch of the microstrip antenna and full ground plane on the bottom of the substrate uses copper with thickness of 0.035 mm. This work is intended to design rectangular single patch antenna with resonant frequency of 3.2 GHz. The initial dimensions of the antenna from the calculation is shown in Table 1. This initial size is then optimized through simulation to get optimal performance before the antenna is being fabricated.

The simulations are conducted using Advanced Design System (ADS) 2019 software at range frequency of 3 - 3.4 GHz with the mesh resolution of 120 cell/wavelength. For optimization, the dimensions of the patch antenna are adjusted to get optimum results of antenna characteristics. Figure 2 depicts top view of the final proposed antenna geometry. The size of the patch antenna is 43×25 mm. The detailed size of each segment of final designed antenna are enlisted in Table 2.

TABLE 1				
INITIAL DIMENSIONS OF DESIGNED ANTENNA				
Parameter	Dimensions (mm)			
Patch size $(W \times L)$	33.15×27.06			
Ground plane size $(W_g \times L_g)$	33.29 × 27.21			

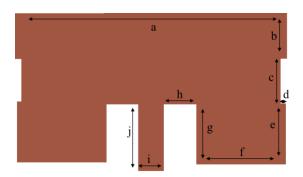


Figure 2. Simulated antenna geometry

TABLE 2 Final dimensions of designed antenna				
Parameter	Dimensions (mm)			
a	43			
b	7.5			
с	7.5			
d	1			
e	10			
f	14			
g	10			
h	5.5			
i	4			
j	11			

B. Antenna Realization

The antenna is fabricated using pyralux FR 9111 with dielectric constant, $\varepsilon_r = 3$, and the thickness, h = 0.025 mm. Pyralux is chosen due to its flexibility, lightness, and low cost in manufacturing technique. A rectangular shaped patch antenna is etched on the top of pyralux substrate. The ground plane is placed on the bottom side of the substrate. The shape of the indentation in the design is intended to get the desired bandwidth and working frequency. This prototype is expected work well at S-band, particularly with operating frequency of 3.2 GHz. The prototype of antenna realization is shown in Figure 3.



Figure 3. Realization of conformal microstrip antenna

III. RESULTS AND DISCUSSION

A. Simulation Results

Return loss defines the delivering of the electrical energy from feed point to an antenna. To have full energy transfer, it is requisite to have negative return loss [13]. Figure 4 depicts the graph of return loss (dB) against frequency (GHz). The simulated return loss results are - 35,80 dB at the the center frequency of 3.2 GHz, -10,39 dB at left side frequency of 3.18 GHz, and -10.40 dB for right side frequency of 3.22 GHz. This result indicates that the proposed antenna design can efficiently transmit or receive a signal at the range of 3.18-3.22 GHz, since the return loss is below the commercially accepted return loss of -9.54 dB at this range frequency. The simulation results are outlined in Table 3. The center frequency is 3.2 GHz with the bandwidth of 40 MHz.

B. Measurement Results

After the fabrication of the proposed antenna, the antenna parameters are measured using Anritsu MS46322A Vector Network Analyzer (VNA) that has frequency range up to 40 GHz. The antenna operating frequency is in the range of 3-3.4 GHz. The measurement process was carried out at four different bending angles scenarios including 0°, 30°, 60°, and 90° as shown in Figure 5.

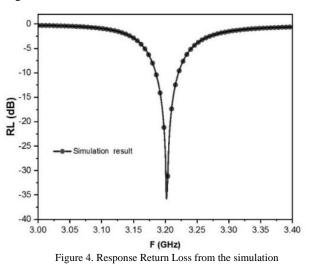
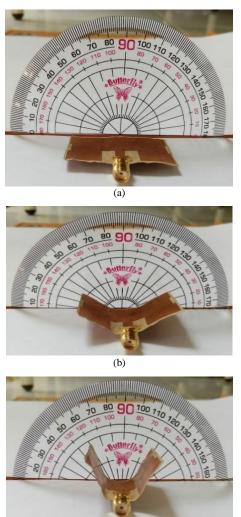


TABLE 3 SIMULATION RESPONSE OF DESIGNED ANTENNA

Specifications	Response
Type Antenna	Conformal
Range Frequency	3 GHz – 3.4 GHz
Substrate	Pyralux FR 9111
f_1 (side left graph)	3.18 GHz
f _c (center frequency)	3.20 GHz
f_2 (side right graph)	3.22 GHz
Return Loss f ₁	-10.39 dB
Return Loss f _c	-35.80 dB
Return Loss f ₂	-10.40 dB
Bandwidth	40 MHz



(c)

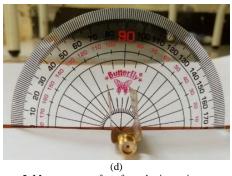


Figure 5. Measurement of conformal microstrip antenna at different bending angles: (a) 0° , (b) 30° , (c) 60° , (d) 90° .

1) Return Loss, Bandwidth, and Impedance

Figure 6 illustrates the comparison of return loss (R_L) from simulation and measurement of the proposed antenna at different bending angles. It can be seen from this Figure that the simulation and the measurement give different results in return loss and center frequency. This difference is possibly caused by different value of dielectric constant (ε_r) and wavelength (λ) of the

substrate, as well as the change in the thickness of the conductor patch layer during the PCB etching process. The simulation result displays 3.20 GHz as center frequency, while the realization center frequency is shifted to higher frequency of 3.28 GHz but still remain the same at various conformal status.

The measurement with various bending angles of 0° , 30° , 60° , and 90° have the same shape and peak frequency. However, the measured return loss for different conformal status has slightly different values. The antenna exhibits the smallest measured return loss of -29.02 dB at the bending angle of 0° and the highest value of -31.38 dB at the bending angle of 90° .

Bandwidth of antenna is defined as the range between upper cut and lower cut of frequency, both at 10 dB, which indicates range frequency where the antenna performs satisfactory. From Figure 6, it can be seen that the center frequency of realized antenna is shifted from 3.20 GHz in simulation to 3.28 GHz for all bending angles scenarios. The bandwidth for measured angles 0°, 60°, and 90° is 80 MHz, while for bending angle of 30°, the measured bandwidth is 90 MHz.

Figure 7 shows antenna impedance characteristics of all bending angles scenarios in the Smith-Chart. It can be seen that the impedance circles for all bending angles have similar impedance value, which are quite close to impedance characteristic of transmission line (50 Ω). As the impedance is close enough to the transmission line impedance, it can be said that the antenna has maximum power transfer which leads to good performance in transmitting and receiving information signal. Table 4 summarizes the measurement results of antenna characteristics. Overall, the best performance was achieved by the scenario of bending angle 90°, in which $R_{\rm L}$ = - 31.38 dB at 3.29 GHz, the bandwidth is 80 MHz, and the impedance $Z_A = 48.36 + j2.04 \Omega$. Even though the simulation and measurement give slightly different results, the parameters can be considered well enough.

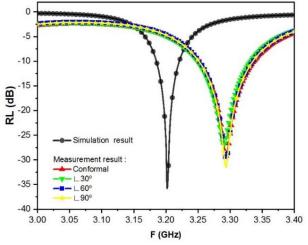


Figure 6. Simulation and measurement result of R_L responses

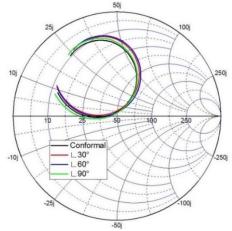


Figure 7. Impedance (Z_A) of conformal microstrip antenna at bending angels of $0^{\circ},\,30^{\circ},\,60^{\circ},\,90^{\circ}$

TABLE 4 MEASUREMENT RESULTS OF ANTENNA CHARACTERISTICS AT FREQUENCY RANGE OF 3 – 3.4 GHZ

Specifications	Response			
Bending angles	0°	30°	60°	90°
f_1	3.25 GHz	3.24 GHz	3.25 GHz	3.25 GHz
f _c	3.29 GHz	3.29 GHz	3.29 GHz	3.29 GHz
f ₂	3.33 GHz	3.33 GHz	3.33 GHz	3.33 GHz
Return Loss f ₁	-10.20dB	-10.04 dB	-10.00	-10.04
			dB	dB
Return Loss f _c	-29.02dB	-26.80 dB	-29.60	-31.38
			dB	dB
Return Loss f2	-10.02dB	-10.09 dB	-10.03	-10.08
			dB	dB
Bandwidth	80 MHz	90 MHz	80 MHz	80 MHz
Impedance (Z _A)	48.37 +j31	48.31 +j4.13	48.42 +j2.7	48.36 +j2.04

2) Radiation Pattern

Figures 8(a)-(d) show measurement results of antenna radiation patterns in various conformal states. Figure 8(a) shows measured radiation patterns at initial state, where the antenna is in flat position with bending angle of 0°. The best signal reception level is achieved at the angle of 0° with the value of -69 dBm. The smallest level of signal reception is -92.87 dBm at 230°. For bending angle of 30° as depicted in Figure 8(b), the signal reception level ranges from -86.12 dBm to -74 dBm. The level varies from of -65.49 dBm to -55.4 dBm as shown in Figure 8(c) for conformal state with bending angle of 60° . Figure 8(d) illustrates the measured signal reception levels for conformal state with bending angle of 90°. The smallest level is -86.6 dBm at 170° and the highest level is -70.9 at 0°. The maximum realized gain is 1.4 dBi.

CONCLUSION

A single path conformal microstrip antenna at Sband frequency has been designed, manufactured and measured. Even though the experimental results have slightly difference from the simulation results, the proposed antenna still can operate in the targeted frequency since other parameters that affect the antenna

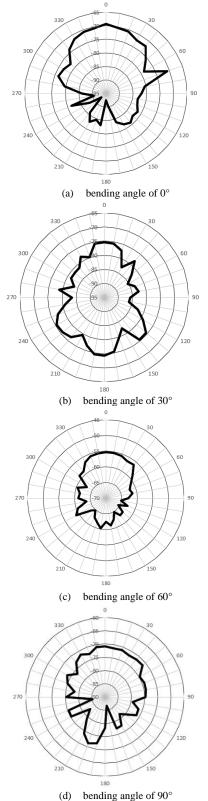


Figure 8. Radiation patterns at various conformal states.

performance are still in an acceptable condition. Operation frequency of the designed conformal patch antenna is 3.2 GHz, with return loss response of 29.02 dB. Antenna measurement at different bending angles scenarios gives quite similar results with a considerable difference in return lost parameter. This indicates that the designed antenna can be used for various communication

purposes in the S-Band frequency and can be etched to the various forms of devices.

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