Design and Realization of FIR Filter for Inter Satellite Link at 50-90 MHZ Frequency using FPGA

Desain dan Realisasi Filter FIR untuk *Inter Satellite Link* pada frekuensi 50-90 MHz menggunakan FPGA

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Abstract

In this paper, design and realization of FIR filter with a bandwidth of 40 MHz at 50-90 MHz frequency has been proposed. The design was destined to be implemented on the Inter Satellite Links (ISL). This kind of filter had been selected due to a need in linear phase responseon the ISL data communication. Equiripple method was used to design the filter becauseof its reliability in minimizing the magnitude errors. The design of this FIR filter was conducted with theoretical calculation and simulation using the R2012b Matlab. For the implementation, FPGA was used with a VHDL as the programming language with a help of Xilinx ISE Design Suite 14.5. Simulation results in Matlab and Simulink indicated that the filter design could be well implemented on ISL at frequency of 50 MHz - 90 MHz with stopband of 60 db. The phase responseresult of the realized design is quite linear so that the filter is suitable for data communication on the ISL.

Keywords : FIR filter, equiripple, FPGA, VHDL.

Abstrak

Dalam tulisan ini, desain dan realisasi filter FIR dengan *bandwidth* 40 MHz pada frekuensi 50 - 90 MHz telah dibuat. Desain dimaksudkan untuk diterapkan pada *Inter Satellite Link* (ISL). Jenis filter FIR dipilih karena kebutuhan dalam respon fase linier pada komunikasi data ISL. Metode *Equiripple* digunakan untuk merancang filter karena kehandalan dalam meminimalisasi kesalahan. Metodologi desain filter FIR ini dimulai dengan perhitungan teoritis dan simulasi menggunakan R2012b Matlab. Untuk realisasinya, FPGA digunakan dengan VHDL sebagai bahasa pemrograman dan dengan bantuan software Xilinx ISE Design Suite 14,5. Hasil simulasi di Matlab dan Simulink menunjukkan bahwa desain filter dapat diimplementasikan pada ISL pada frekuensi 50 MHz - 90 MHz dengan stopband sebesar 60 db. Hasil respon fase pada realisasi cukup linear sehingga filter cocok digunakan untuk komunikasi data pada ISL.

Kata kunci : Filter FIR, equiripple, FPGA, VHDL.

I. INTRODUCTION

The satellite is an object that orbits around the earth that is normally used by humans as to communicated, weather monitoring, etc. The satellite technology continues to evolve. An Inter satellite links (ISL) is one of the technology that can make the satellites connect directly with other satellites.

The ISL is a solution of the existing problems on LEO satellites monitoring the earth. These satellites can only send a picture when the position above the earth station which causes the delay to get information. Therefore, the earth stations is made with large numbers in order to continue monitoring the LEO satellite. However, this requires a considerable high cost. Because of that the ISL which can communicate the satellites directly with other satellites assessed more quickly and efficiently.

Filter is one of the components used in the ISL. The filter is used to pass the signals by passing frequency digital signal which contains the required info and reduce the undesired signal. In the recent years, many research of Finite Impulse Response (FIR) filters had been conducted. Some applications that could be implemented using this kind of filters are: image sampling [1], image coding [2], beamforming [3], software radio [4], audio [5], hearing aid [6] and DSP application [7]

In this paper, design and realization of FIR Filter with bandwidth of 40 MHz at 50-90 MHz frequency for ISL application are done. An Equiripple method was used to design the filter because of its reliability in minimizing the magnitude errors. The design of this FIR filter was conducted with theoretical calculations and simulated using Matlab R2012b. For the implementation, FPGA was used with VHDL as the

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programming language and a help of Xilinx ISE Design Suite 14.5

II. FIR FILTER

A filter is one of the component in communication systems that is used to pass the desired frequency and stop unwanted frequency [8]. Band Pass Filter is one of the types of filters that has characters to allowed frequency in the range of a particular region (passband), and reduce the frequency area above and below the frequency range (stopband) [8].

The FIR filter is one type of filters which had a non recursive characteristic of the current output. It means that characteristic of the current output does not depend on the previous output, so it has a limited impulse response [8]. This makes the output of filter is calculated as the value of sum finite figures in the past, present, and possible future values of the input filter. Formulated in the Equation 1 [8] :

$$y[n] = \sum_{k=0}^{M} b_{k X[n-k]} = \sum_{k=0}^{M} h(k) \cdot X[n-k]$$
(1)

FIR filterhas the following transfer function in Equation 2 [9]:

$$H(z) = \frac{Y(z)}{X(z)} = \frac{\sum_{k=0}^{M} b_{k \ Z^{-k}]}}{1 + \sum_{k=1}^{N} a_{k \ Z^{-k}]}}$$
(2)

The impulse response of the ideal filter is showed in Equation 3[10]:

$$h(n) = \frac{1}{2\pi} \int_{-\omega c2}^{-\omega c1} e^{j\omega n} d\omega + \int_{\omega c1}^{\omega c2} e^{j\omega n} d\omega$$
$$= \frac{1}{\pi n} \sin(\omega c2n) - \frac{1}{\pi n} \sin(\omega c1n)$$
$$h(n) = \frac{1}{2\pi} \int_{-\omega c2}^{-\omega c1} e^{j\omega n} d\omega + \int_{\omega c1}^{\omega c2} e^{j\omega n} d\omega$$
$$= \frac{1}{\pi n} \sin(\omega c2n) - \frac{1}{\pi n} \sin(\omega c1n)$$
(3)

The method of sampling frequency and windowing method is the simplest method to design an FIR filter. But both methods it's also has a weakness in the lack of reliability to set the critical frequency as ω_s and ω_p . Equiripple is a method that provides optimization in FIR filter design, to approach the magnitude of the error between the desired frequency response with the actual frequency response that spreads evenly on the passband and stopband so as to minimize the magnitude of the error.

The magnitude of the error can be formulated into Equation 4 [9]:

$$E(e^{j\omega}) = \widehat{W}(e^{j\omega})[\widehat{H}_{dr}(e^{j\omega}) - P(e^{j\omega})]$$
(4)

The weighting function and the desired frequency response is written in Equation 5 [2]:

$$\widehat{W}(e^{j\omega}) = W(e^{j\omega})Q(e^{j\omega})$$
(5)

$$\widehat{H}_{dr}\left(e^{j\omega}\right) = \frac{H_{dr}\left(e^{j\omega}\right)}{O(e^{j\omega})} \tag{6}$$

From the error function $(e^{j\omega})$, the problem of Chebyshev approximation method (minimax) is to seek α_k filter parameters that minimize the maximum

e-ISSN: 2527-9955 p-ISSN: 1411-8289 error of the band frequencydesired which can be formulated with in Equation 7 [9].

$$\min_{over \{\alpha_k\}} \left[\max_{\omega \in S} |E(e^{j\omega})| \right] \\
= \min_{over \{\alpha_k\}} \left[\max_{\omega \in S} \left| \widehat{W}(e^{j\omega}) [\widehat{H}_{dr}(e^{j\omega}) - \sum_{k=0}^{L} \alpha_k \cos \omega k] \right| \right]$$
(7)

III. DESIGN AND SYSTEM REALIZATION

The block diagram of a FIR filter design could be seen in Figure 1.

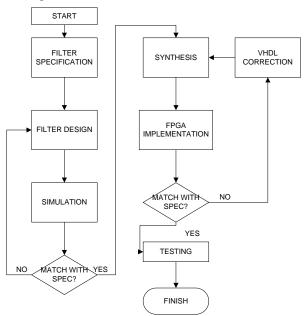


Figure 1. Block Diagram of FIR Filter Design.

1) Matlab Design

The specification of FIR filter design :

| • | Stopband Frequency1 | : 50 MHz |
|----|-------------------------------|----------------|
| • | Cutoff Frequency 1 | : 52 MHz |
| • | Cutoff Frequency 2 | : 88 MHz |
| • | Stopband Frequency 2 | : 90 MHz |
| • | Ripple | : 0,1 dB |
| • | Attenuation | : 60 dB |
| • | Sampling Frequency | : 200MHz |
| In | the method there is a toolhow | called EDA too |

In the matlab, there is a toolbox called FDA tool (Figure 2). This toolbox will help to design a filter.

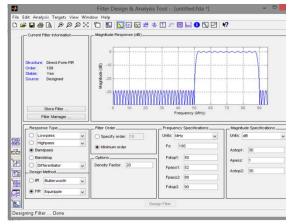


Figure 2. FDA Toolbox.

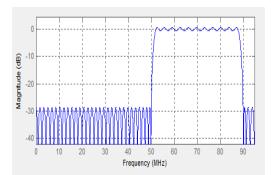


Figure 3. Magnitude Response.

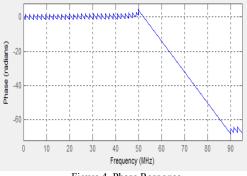
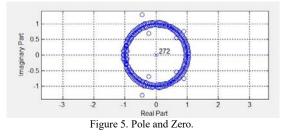


Figure 4. Phase Response.

FDA tool can show the magnitude response(Figure 3), the phase response (Figure 4) and the location of poles and zeros (Figure 5). It can prove whether the phase response of the filter is linear. The layout of the poles and zeros is to prove the stability of filter. Fir filter does not have a zero, because there is no feedback coefficient in it. FIR filter have a feed forward coefficient. This is why the FIR filter can be stabilized. If there is a pole in the loop, filter is stable, contrastly if a pole outside the circle then filter unstable. Total pole will be a number of filter coefficients.



IV. SIMULATION AND ANALYSIS

A. Matlab Analysis

In this scenario, Matlab simulation is conducted with a scale of 1 : 1.000.000. The block diagram of the simulation could be seen in Figure 6. In the simulation, FIR filter will pass a sinusoid signal with a frequency of 45 MHz, 50 MHz, 60 MHz, 65 MHz, 70 MHz, 75 MHz, 80 MHz, 85 MHz, 90 MHz, and 95 MHz. Sinusoid signals with these frequencies will then be passed to digital FIR filter that has been scaled to 1: 1.000.000. Afterwards, the results could be obtained by plotting the filter output of sinusoid signal using Matlab.



Figure 6. Block Diagram of FIR Filter Simulation in Matlab.

Figure 7 represents the output of filtered signals at a frequency of 45 MHz. Input signal and output signal at a frequency of 55 MHz was depicted by Figure 8. The signal contained in the passband region will be passed. But there is a delay of 130 us before the signal can be read by the system. It is caused by a d elay element contained in the filter.

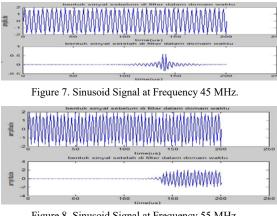


Figure 8. Sinusoid Signal at Frequency 55 MHz.

After a few experiment, magnitude response could be seen in Figure 9 while the phase response is illustrated in Figure 10.

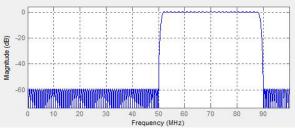
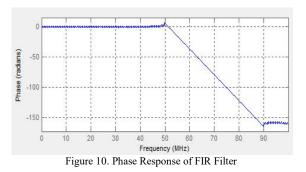


Figure 9. Magnitude Response of FIR Filter.



B. Simulink Analysis

Block Diagram of FIR Filter Simulation in Simulink was depicted in Figure 11. Two sinusoid input signals are generated. The sine signal from the DSP is signal information, while the random source is noise signal. The type of designed filter is a band stop filter which filters output signals originating from a random source, so that the noise signal could not passed the system. Add block serves as a modulator to combine both the information signal and noise signal. Time scope in this block diagram serves to look at three things : Information signal, adder output signal before being filtered and the filter output signal.

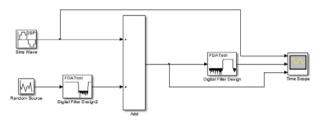


Figure 11. Block Diagram of FIR Filter Simulation in Simulink.

A DSP sine wave frequency sinusoid signal issued by 45 MHz with amplitude equal to 1 could be seen in Figure 12. Random source will emit a signal indeterminate amplitude of 0 - 4 with a frequency between 0-50 MHz and > 90 MHz. The black line graph is sinusoidal signal that will be filtered. Purple line graph is a sinusoid signal that is passed filter. Blue line graph is information signal. The filter did not pass the sinusoid signal. Because the signal below 52 MHz frequency signal will be attenuated.

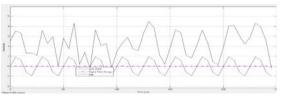


Figure 12. Signal at frequency 45 MHz.

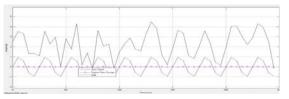


Figure 13. Signal at frequency 55 MHz.

In Figure 13, it could be seen a DSP sine wave frequency sinusoidal signal issued by 55 MHz with an amplitude equal to 1. Random source will emit a signal indeterminate amplitude of 0 - 4 with a frequency between 0-50 MHz and > 90 MHz. The filter passed a sinusoid signal which has the same amplitude but also has a slightly different form. This is due to the delay of 130 μs in the FIR filter derived from the components forming the FIR filter.

From the Table 1, we can concluded that the digital filter could be designed and implemented on the FPGA device, because FPGA resources utilization were below 100 %.

C. Designed Filter Performance Analysis

Block diagram scenario of implemented FIR filter on FPGA could be seen in Figure 14. FPGA ability to generate the signal was used in this scenario. The output digital FIR filter could be viewed using chipscope as logic analyzer. Then the data will be collected and converted into decimal.

FPGA RESOURCES FOR IMPLEMENTATION

| Device Utilization Summary (estimated values) | | | | |
|---|-------|-----------|-------------|--|
| Logic Utilization | Used | Available | Utilization | |
| Number of Slice Registers | 3860 | 28800 | 13% | |
| Number of Slice LUTs | 11382 | 28800 | 39% | |
| Number of fully used LUT- FF pairs | 691 | 14551 | 4% | |
| Number of bonded IOBs | 19 | 480 | 3% | |
| Number of Block RAM/FIFO | 1 | 60 | 1% | |
| Number of BUFG/BUFGCTRLs | 2 | 32 | 6% | |
| Number of DSP48Es | 45 | 48 | 93% | |
| Number of PLL_ADVs | 1 | 6 | 16% | |
| Average Fanout of Non- Clock Nets | 2.73 | | | |

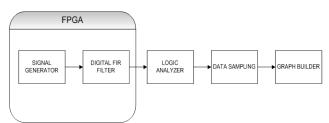
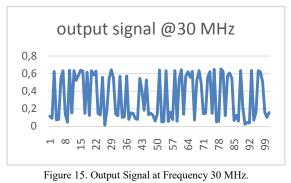


Figure 14. Block Diagram of FIR Filter Implementation in FPGA.

FPGA ability to generate the signal was used in this scenario. The output digital FIR filter could be viewed using chipscope as logic analyzer. Then the data will be collected into a table and will be converted into decimal. Thus the output of the digital filter could be seen to builda magnitude response graphic. The data that taken from chipscope sampled 100 time. From these tests the output waveform obtained as seen in Figure 15. The output signal of the filter with a frequency of 30 MHz or below the pass band has a shape signal and did not pass sinusoid wave form. It was caused by the signal at that frequency attenuated by the system so it made the shape signal becomes damaged.



In Figure 16, the output signal with a frequency of 70 MHz forming a sinusoid wave could be seen. This happens because the frequency of the signal passed by the system that makes the shape look like a sinusoid wave signal. Magnitude response of the output signal with a frequency of 30 MHz, 40 MHz, 50 MHz, 60

 $\rm MHz$, 70 $\rm MHz$, 80 $\rm MHz$, 90 $\rm MHz$, and 100 $\rm MHz$ could be seen in Figure 17.

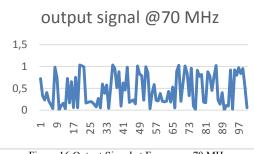


Figure 16.Output Signal at Frequency 70 MHz.

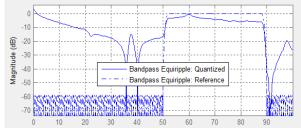


Figure 17. Magnitude response of Filter.

In Figure 17, there are two lines. The dashed line is the result of simulation and solid lines are the result of implementation. From the result, frequency response of the filter implementation has been good. Compared with the results of the simulation, bandwidth implementation results widened to 40 - 90 MHz. This happens due to several factor, such as the size of the sampling frequency used resulting order is not enough to make the filter bandwidth of 50 - 90 MHz.

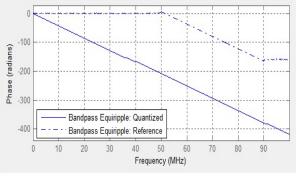


Figure 18. Phase response of Filter

In Figure 18 the results of the simulation and implementation of the system is fairly linear phase response at the desired frequency range 50 - 90 MHz, this proves that the FIR filter is capable of making a linear phase response and is suitable for data communication.

CONCLUSION

In this paper, the design and realization of FIR Filter with a bandwidth of 40 MHz at 50 - 90 MHz frequency has been done.Compared with the results of the simulation, bandwidth results of the implemented filter widened to 40 - 90 MHz due to several factors such as the size of the sampling frequency used

resulting order is not enough to make the filter bandwidth of 50 - 90 MHz. Although overall, based on the results from simulation and implementation, it can be concluded that digital FIR filter designed with equiripple method can be suitable for data communication in ISL application.

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